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# **Phase reconstruction of spectrograms with linear unwrapping : application to audio signal restoration**

*Reconstruction de phases de spectrogrammes par déroulé linéaire : application à la restauration de signaux audio*



avril 2015

Département Traitement du Signal et des Images Groupe AAO : Audio, Acoustique et Ondes

# Phase re
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#### Abstra
t

This paper introduces a novel technique for reconstructing the phase of modified spectrograms of audio signals. From the analysis of mixtures of sinusoids we obtain relationships between phases of successive time frames in the Time-Frequency (TF) domain. To obtain similar relationships over frequencies, in particular within onset frames, we study an impulse model. Instantaneous frequencies and attack times are estimated locally to encompass the class of non-stationary signals such as vibratos. These techniques ensure both the verti
al oheren
e of partials (over frequen
ies) and the horizontal oheren
e (over time). The method is tested on a variety of data and demonstrates better performance than traditional consistency-based approa
hes. We also introdu
e an audio restoration framework and observe that our te
hnique outperforms traditional methods.

### Key words

Phase reconstruction, sinusoidal modeling, linear unwrapping, phase consistency, audio restoration.

### Résumé

Ce rapport présente une nouvelle technique pour la reconstruction de phases de spectrogrammes modifiés. A partir de l'analyse de mélanges de sinusoïdes, on obtient des relations entre les phases trames successives dans le plan temps-fréquence (TF). Pour obtenir des relations similaires entre fréquences, en particulier au sein des trames d'attaque, nous étudions un modèle d'impulsion. Les fréquen
es instantanées et les temps d'attaque sont estimés localement afin de pouvoir représenter des signaux non stationnaires, tels que les vibratos. Ces te
hniques permettent d'assurer à la fois une ohéren
e verti
ale entre les partiels (à travers les fréquen
es) et horizontale (au ours du temps). Cette méthode est testée sur des données expérimentales, et montre de meilleurs résultats que l'appro
he traditionnelle basée sur la onsistan
e. Nous proposons également d'introduire ette te
hnique dans un ontexte de restauration de signaux audio, dans lequel une meilleure performan
e qu'ave les méthodes traditionnelles est observée.

#### Mots lés

Re
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tion de phase, mélanges de sinusoïdes, déeroulé linéaire, onsistan
e de phase, restauration de signaux audio

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## 1 Introdu
tion

A variety of music signal processing techniques act in the TF domain, exploiting the particular structure of music signals. For instance, the family of techniques based on Nonnegative Matrix Factorization (NMF) is often applied to spectrogram-like representations, and has proved to provide a successful and promising framework for source separation [1]. Magnitude-recovery techniques are also useful for restoring missing data in corrupted signals  $[2]$ .

However, when it omes to resynthesize time signals, the phase re
overy of the orresponding Short-Time Fourier Transform (STFT) is necessary. In the source separation framework, a common practice consists in applying Wiener-like filtering (soft masking of the complex-valued STFT of the original mixture). When there is no prior on the phase of a component (e.g. in the context of audio restoration), a consistency-based approach is often used for phase recovery [3]. That is, a complex-valued matrix is iteratively computed to be close to the STFT of a time signal. A recent benchmark has been conducted to assess the potential of source separation methods with phase recovery in NMF [4]. It points out that consistency-based approaches provide poor results in terms of audio quality. Besides, Wiener filtering fails to provide good results when sources overlap in the TF domain. Thus, phase re
overy of modied audio spe
trograms is still an open issue. The High Resolution NMF (HRNMF) model [5] has shown to be a promising approach, since it models a TF mixture as a sum of autoregressive (AR) components in the TF domain, thus dealing explicitly with a phase model.

Another approach to reconstruct the phase of a spectrogram is to use a phase model based on the observation of fundamental signals that are mixtures of sinusoids. Contrary to consistency-based approaches using the redundancy of the STFT, this model exploits the natural relationship between adjacent TF bins due to the model. This approach is used in the phase vocoder algorithm [6], although it is mainly dedicated to time stretching and pitch modification of signals, and it requires the phase of the original STFT. More recently, [7] proposed a omplex NMF framework with phase onstraints based on sinusoidal modeling. Although promising, this approach is limited to harmonic and stationary signals, and requires prior knowledge on fundamental frequen
ies and numbers of partials.

In this paper, we propose a generalization of this approach that consists in estimating the phase field of mixtures of sinusoids from its explicit calculation. We then obtain an algorithm which unwraps the phases horizontally (over time frames) to ensure the temporal coherence of the signal, and vertically (over frequency hannels) to enfor
e spe
tral oheren
e between partials, whi
h are naturally observed in musi
al a
ousti
s. Our technique is suitable for a variety of pitched music signals, such as piano or guitar sounds. A dynamic estimation (at ea
h time frame) of instantaneous frequen
ies extends the validity of this te
hnique to non-stationary signals such as cellos and speech. This technique is tested on a variety of signals and integrated in an audio restoration framework.

The paper is organized as follows. Section 2 presents the horizontal phase unwrapping model. Section 3 is dedicated to phase reconstruction on onset frames. Section 4 presents a performance evaluation of this technique through various experiments. Se
tion 5 introdu
es an audio restoration framework using this phase re
overy method. Finally, section 6 draws some concluding remarks.

# 2 Horizontal phase reconstruction

### 2.1 Sinusoidal modeling

Let us consider a sinusoid of normalized frequency  $f_0 \in [-\frac{1}{2};\frac{1}{2}]$ , origin phase  $\phi_0 \in [-\pi;\pi]$  and amplitude  $A > 0$ :

$$
\forall n \in \mathbb{Z}, \ x(n) = A e^{2i\pi f_0 n + i\phi_0}.\tag{1}
$$

The expression of the STFT is, for each frequency channel  $k \in \llbracket -\frac{F-1}{2}; \frac{F-1}{2} \rrbracket$  (with F the odd-valued Fourier transform length) and time frame  $t \in \mathbb{Z}$ :

$$
X(k,t) = \sum_{n=0}^{N-1} x(n+tS)w(n)e^{-2i\pi \frac{k}{F}n}
$$
\n(2)

where w is a N sample-long analysis window and S is the time shift (in samples) between successive frames. Let  $W(f) = \sum_{n=0}^{N-1} w(n) e^{-2i\pi f n}$  be the discrete time Fourier transform of the analysis window for each normalized frequency  $f \in [-\frac{1}{2}; \frac{1}{2}]$ . Then the STFT of the sinusoid (1) is:

$$
X(k,t) = Ae^{2i\pi f_0 St + i\phi_0} W\left(\frac{k}{F} - f_0\right). \tag{3}
$$

The unwrapped phase of the STFT is then:

$$
\phi(k,t) = \phi_0 + 2\pi S f_0 t + \angle W\left(\frac{k}{F} - f_0\right)
$$
\n<sup>(4)</sup>

where  $\angle z$  denotes the argument of the complex number z. This leads to a relationship between two successive time frames:

$$
\phi(k, t) = \phi(k, t - 1) + 2\pi S f_0.
$$
\n(5)

More generally, we can compute the phase of the STFT of a frequency-modulated sinusoid. If the frequency variation is low between two successive time frames, we can generalize the previous equation:

$$
\phi(k,t) = \phi(k,t-1) + 2\pi S f_0(t).
$$
\n(6)

Instantaneous frequency must then be estimated at each time frame to encompass variable frequency signals such as vibratos, which commonly occur in music signals (singing voice or cello signals for instance).

#### Instantaneous frequency estimation 2.2

Quadratic interpolation FFT (QIFFT) is a powerful tool for estimating the instantaneous frequency near a magnitude peak in the spectrum [8]. It consists in approximating the shape of a spectrum near a magnitude peak by a parabola. This parabolic approximation is justified theoretically for Gaussian analysis windows, and used in practical applications for any window type. The computation of the maximum of the parabola leads to the instantaneous frequency estimate. Note that this technique is suitable for signals where only one sinusoid is active per frequency channel.

The frequency bias of this method can be reduced by increasing the zero-padding factor [9]. For a Hann window without zero-padding, the frequency estimation error is less than 1 %, which is hardly perceptible in most music applications according to the authors.

### 2.3 Regions of influence

When the mixture is composed of one sinusoid, the phase must be unwrapped in all frequency channels according to (5) using the instantaneous frequency  $f_0$ . When there is more than one sinusoid, frequency estimation is performed near each magnitude peak. Then, the whole frequency range must be decomposed in several regions (regions of influence  $[6]$ ) to ensure that the phase in a given frequency channel is unwrapped with the appropriate instantaneous frequen
y.

At time frame t, we consider a magnitude peak  $A_p$  in channel  $k_p$ . The magnitudes (resp. the frequency channels) of neighboring peaks are denoted  $A_{p-1}$  and  $A_{p+1}$  (resp.  $k_{p-1}$  and  $k_{p+1}$ ). We define the region of influence  $I_p$  of the p-th peak as follows:

$$
I_p = \left[ \frac{A_p k_{p-1} + A_{p-1} k_p}{A_p + A_{p-1}}; \frac{A_p k_{p+1} + A_{p+1} k_p}{A_p + A_{p+1}} \right].
$$
\n(7)

The greater  $A_p$  is relatively to  $A_{p-1}$  and  $A_{p+1}$ , the wider  $I_p$  is. Note that other definitions of regions of influence exist, such as choosing the limit between two peaks as the channel of lowest energy [6].

## 3 Onset phase re
onstru
tion

#### $3.1$ Impulse model

Impulse signals are useful to obtain a relationship between phases over frequencies (vertical unwrapping) [10]. Although they do not accurately model attack sounds, they provide simple equations that can be further improved for more omplex signals. The model is:

$$
\forall n \in \mathbb{Z}, \, x(n) = A\delta_{n-n_0} \tag{8}
$$

where  $\delta$  is equal to one if  $n = n_0$  (the so-called *attack time*) and zero elsewhere and  $A > 0$  is the amplitude. Its STFT is equal to zero except within attack frames:

$$
X(k,t) = Aw(n_0 - St)e^{-2i\pi \frac{k}{F}(n_0 - St)}.
$$
\n(9)

We can then obtain a relationship between the phases of two successive frequency channels within an onset frame, assuming that  $w > 0$ :

$$
\phi(k,t) = \phi(k-1,t) - \frac{2\pi}{F}(n_0 - St).
$$
\n(10)

The similarity between (10) and (5) was expected because the impulse is the dual of the sinusoid in the TF domain. This comparison naturally leads to estimating parameter  $n_0$  (the "instantaneous" attack time) in each frequency channel as we previously estimated  $f_0$  (the instantaneous frequency) in each time frame (cf. equation (6)). This leads to the following verti
al unwrapping equation:

$$
\phi(k,t) = \phi(k-1,t) - \frac{2\pi}{F}(n_0(k) - St).
$$
\n(11)

### 3.2 Attack time estimation

In order to estimate  $n_0(k)$ , we look at the magnitude of the STFT of the impulse in a frequency channel k:

$$
|X(k,t)| = Aw(n_0(k) - St).
$$
\n(12)

We then choose  $n_0$  such that the STFT magnitude of the impulse over onset frames has a shape similar to that of the analysis window. For instan
e, a least-squares estimation method an be used. Syntheti mixtures of impulses are perfectly reconstructed with this technique. Alternatively, we can also estimate  $n_0(k)$  with a temporal QIFFT and update the phase with (11).

## 4 Experimental evaluation

#### Protocol and datasets 4.1

The MATLAB Tempogram Toolbox [11] provides a fast and reliable onset frames detection from spectrograms. We use several datasets in our experiments:

- A: 30 mixtures of piano notes from the Midi Aligned Piano Sounds (MAPS) database [12],
- B: 30 piano pie
es from the MAPS database,
- C: 12 string quartets from the SCore Informed Source Separation DataBase (SCISSDB) [13],
- D: 40 speech excerpts from the Computational Hearing in Multisource Environments (CHiME) database [14].

The data is sampled at  $F_s = 11025$  Hz and the STFT is computed with a 512 sample-long Hann window and 75 % overlap. The Signal to Distortion Ratio (SDR) is used for performance measurement. It is computed with the **BSS** Eval toolbox  $[15]$  and expressed in dB. The popular consistency-based Griffin and Lim  $(GL)$ algorithm [3] is also used as a reference. We run 200 iterations of this algorithm (performance is not further improved beyond). It is initialized with known phase values if any or random values if not, and results are averaged on 30 initializations. Simulations are run on <sup>a</sup> 3.60GHz CPU pro
essor and 16Go RAM omputer.



Figure 4.1: Spectrogram of a mixture with vibrato (left) and instantaneous frequencies in the 2800 Hz channel (right)

Dataset	Error	GL	PU
	0.38	6.9	$2.5\,$
В	0.36	-12.6	$1.7\,$
	0.41		5.3
	0.52		0.5

Table 1: Frequency estimation error (%) and reconstruction performance (SDR in dB) for various audio datasets

#### $4.2$ Horizontal phase reconstruction

A first experiment consists in estimating instantaneous frequencies on synthetic mixtures of damped sinusoids, which parameters (in particular the frequencies) are user-defined. Frequency estimation error with QIFFT is below the threshold of 0.2 %, ommonly referred to as the maximal human auditory resolution.

Figure 4.1 illustrates the instantaneous frequencies estimated with the phase vocoder technique [6], used as a reference, and with our algorithm on a vibrato. Identical results are obtained. Our method is thus suitable for estimating variable instantaneous frequency signals as well as stationary components. We computed the average frequency error between phase vocoder and QIFFT estimates for the datasets presented in section 4.1. The results presented in the first column of Table 1 confirm that QIFFT provides an accurate frequency estimation.

Table 1 also presents reconstruction performance (assuming onset phases are known) for both Griffin and Lim (GL) and our Phase Unwrapping (PU) algorithms. Our approach significantly outperforms the traditional GL method: both stationary and variable frequency signals are reconstructed accurately. In addition, our algorithm is faster than the GL te
hnique: on a 3min 48<sup>s</sup> piano pie
e, the re
onstru
tion is performed in 18<sup>s</sup> with our approach and in 623s with GL algorithm.

### 4.3 Onset phase re
onstru
tion

Onset phases can be reconstructed with  $n_0$ -estimation using the impulse magnitude (PU-Impulse) or with QIFFT (PU-QIFFT). We also test random phases values (PU-Rand, no vertical coherence), zero phases (PU-0, partials in phase) and alternating partial phases between 0 and  $\pi$  (PU-Alt, phase-opposed partials). These choices are justified by the observation of the relationship between partials in musical acoustics [16]. The phase of the partials is then fully re
overed with horizontal unwrapping. We test these methods on dataset A. Results presented in Table 2 show that all our approaches provide better results than GL algorithm on this class of signals. Onset phase unwrapping with  $n_0$ -estimation based on QIFFT provides the best result, ensuring some form of vertical coherence. In particular, we perceptually observe that this approach provides a neat percussive atta
k.

Method	$SDR$ (dB)
GL.	-7.9
PU-Impulse	$-4.0$
PU-QIFFT	$-2.6$
PU-Rand	$-4.3$
$P U - 0$	4.7
$PU-Alt$	-3.5

Table 2: Signal reconstruction performance of different methods on dataset A



Figure 4.2: Reconstruction performance of different methods and percentages of corruption on dataset A

#### Complete phase reconstruction  $4.4$

We consider unaltered magnitude spectrograms from dataset A. A variable percentage of the STFT phases is randomly corrupted. We evaluate the performance of our algorithm to restore the phase both on onset and non-onset frames.

Figure 4.2 confirms the potential of this technique. Our method produced an average increase in SDR of 6dB over the orrupted data. It also performs better than the GL algorithm when a high per
entage of the STFT phases must be re
overed.

However, note that this experiment consists in phase reconstruction of *consistent* spectrograms (*i.e* positive matri
es that are the magnitude of the STFT of a time signal): GL algorithm is then naturally advantaged in this case. Realistic applications (cf. next section) involve the restoration of both phase and magnitude, which leads to in
onsistent spe
trograms.

The goal of audio inpainting is to restore corrupted or missing value of a signal. Since corruption can be done in the temporal domain or in the TF domain, we will study those two approa
hes.

First, we will partially recover phase from a consistent spectrogram (the STFT magnitude is not modified). Secondly, we will reconstruct whole parts of STFT (both magnitude and spectrogram). Finally, a time signal is corrupted with clicks and we compare restoration with a temporal method, our approach, and HR NMF algorithm  $[5]$ .

## 5 Appli
ation of phase re
onstru
tion to audio restoration

%subse
tionTemporal orruption: li
k removal

A common alteration of music signals is the presence of noise on short time periods (a few samples) called clicks. We corrupt time signals with clicks that represent less than 1 % of the total duration. Clicks are obtained by differentiating a 10 sample-long Hann window.

Magnitude restoration of missing bins is performed by linear interpolation of the log-magnitudes in ea
h frequency channel. Figure 5.1 illustrates this technique. Phase recovery is then performed with our method



Figure 5.1: Restoration of spe
trogram by linear interpolation of the log-magnitudes on a piano notes

Dataset	ΑR	<b>HRNMF</b>	GL	PU
		16.9	8.6	11.7
	4.3	10.9	5.9	$\bf 7.1$
	8.2	10.6	$6.6\,$	7.1
	8.3	10.9	Q	9.4

Table 3: Signal restoration performan
e (SDR in dB) for various methods and datasets

(PU) or alternatively with the GL algorithm. We ompare those results to the traditional restoration method based on autoregressive  $(AR)$  modeling of the time signal [18], and with HRNMF [5].

Table 3 presents results of restoration. HRNMF provides the best results in terms of SDR. Though, our approa
h outperforms the traditional method and GL algorithm. Besides, we underline that the HRNMF model uses the phase of the nonorrupted bins, while our algorithm is blind. Lastly, our te
hnique remains faster than HRNMF: for a 3min55<sup>s</sup> piano pie
e, restoration is performed in 99<sup>s</sup> with our algorithm and in 222<sup>s</sup> with HRNMF.

# 6 Con
lusion

The new phase reconstruction technique introduced in this work appears to be an efficient and promising method. The analysis of mixtures of sinusoids leads to relationships between successive TF bins phases. Physical parameters su
h as instantaneous frequen
ies and atta
k times are estimated dynami
ally, en
ompassing a variety of signals such as piano and cellos sounds. The phase is then unwrapped in all frequency channels for onset frames and over time for partials. Experiments have demonstrated the accuracy of the unwrapping method, and we integrated it in an audio restoration framework. Better results than with traditional methods have been rea
hed.

The re
onstru
tion of onset frames still needs to be improved as suggested by the variety of data. Further work will focus on exploiting known phase data for reconstruction: missing bins can be inferred from observed phase values. Alternatively, time-invariant parameters such as phase offsets between partials [19] can be used. Su
h developments will be introdu
ed in an audio sour
e separation framework, where the phase of the mixture an be exploited.

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