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## A Challenge Code for Maximizing the Entropy of PUF Responses

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## Outline

Entropy of PUF
Concept
Estimation
Prevision
Running example: the Loop-PUF (LPUF)
SRAM PUF example
L-PUF into details [CDGB12]
Theory
Definitions
Results
Main result
Beyond $n$ bits
Conclusions

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## Entropy of PUFs



PUFs are instanciations of blueprints by a fab plant

## After fabrication

(a)


## (estimation $\hat{\mathbb{P}}$ )

(b)


Which PUF is the most entropic?

## (estimation $\hat{\mathbb{P}}$ )

(a)

(b)


Which PUF is the most entropic?

$$
\text { Recall } H=-\sum_{c=0 \times 00 \ldots 00}^{0 x f f \ldots f f} \mathbb{P}(R=\operatorname{PUF}(c)) \log \mathbb{P}(R=\operatorname{PUF}(c))
$$

## Before fabrication

## Stochastic model

- Active discussion at ISO sub-committee 27:


ISO/IEC JTC 1/SC 27/WG 3 N1 233

REPLACES:

| ISO/IEC JTC 1/SC 27/WG 3 |
| :---: |
| Information technology - Security techniques - Security evaluation, testing and specification |
| Convenorship: AENOR, Spain, Vice-convenorship: JISC, Japan |

DOC TYPE: working draft

TITLE: $\quad$ Text for ISO/IEC 1st WD 20897 - Information technology — Security requirements and test methods for physically unclonable functions for generating non-stored security parameters

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## Non-delay PUF: SRAM PUF



Amount of entropy: $=n$.

## Delay PUF: core delay element



Same idea as in other delay PUFs, like arbiter-PUF, etc.

Let $d\left(c_{i}\right)$ be the corresponding delay. As time is an extensive physical quantity:

$$
d\left(c_{i}\right)= \begin{cases}d_{i}^{T_{1}}+d_{i}^{B_{2}}=d_{i}^{T_{B}} & \text { if } c_{i}=-1, \\ d_{i}^{B_{1}}+d_{i}^{T_{2}}=d_{i}^{B T} & \text { if } c_{i}=+1 .\end{cases}
$$

The delays $d_{i}^{T B}$ and $d_{i}^{B T}$ are modeled as i.i.d. normal random variables selected at fabrication [PDW89].


Figure: Monte-Carlo simulation (with 500 runs) of the delays in a chain of 60 basic buffers implemented in a 55 nm CMOS technology.

## Delay PUF: Loop PUF



Amount of entropy: $>n$ ?
Nota bene: here, $d(c)$ is expressed in number of clock cycles.

## LPUF is not self-contained

It needs a protocole

input : Challenge $c$
output: Response $B_{C}$
1 Set challenge $c$
2 Measure $d_{1} \leftarrow\left\lfloor N \sum_{i=1}^{n} d\left(c_{i}\right)\right\rfloor$
3 Set challenge - $c$
4 Measure $d_{2} \leftarrow\left\lfloor N \sum_{i=1}^{n} d\left(-c_{i}\right)\right\rfloor$
5 return $B_{c}=\operatorname{sign}\left(d_{1}-d_{2}\right)$
Algorithm 1: Protocole to get one bit out LPUF.

## RAM-PUF vs Loop-PUF

$$
\text { For } n=8
$$

SRAM-PUF

$(0 \leq M \leq n)$

LPUF

$\left(0 \leq M \leq 2^{n-1}\right)$

SECUREIC

TELECOM ParisTech

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## Challenge

## Definition

A challenge $c$ is a vector of $n$ control bits $c=\left(c_{1}, c_{2}, \ldots, c_{n}\right) \in\{ \pm 1\}^{n}$. Let $\Delta_{1}, \Delta_{2}, \ldots, \Delta_{n}$ be i.i.d. zero-mean normal (Gaussian) variables characterizing the technological dispersion. A bit response to challenge $c$ is defined as

$$
\begin{equation*}
B_{C}=\operatorname{sign}\left(\Delta_{C}\right) \in\{ \pm 1\} \tag{1}
\end{equation*}
$$

where

$$
\begin{equation*}
\Delta_{c}=c_{1} \Delta_{1}+c_{2} \Delta_{2}+\cdots+c_{n} \Delta_{n} . \tag{2}
\end{equation*}
$$

## Challenge code

## Definition

A challenge code $\mathcal{C}$ is a set of $M n$-bit challenges that form a ( $n, M$ ) binary code. We shall identify $\mathcal{C}$ with the $M \times n$ matrix of $\pm 1$ 's whose lines are the challenges.
The $M$ codewords and their complements are used to challenge the PUF elements. The corresponding identifier is the $M$-bit vector

$$
\begin{equation*}
B=\left(B_{C}\right)_{c \in \mathcal{C}} . \tag{3}
\end{equation*}
$$

The entropy of the PUF responses is denoted by $H=H(B)$.

## Orthant probabilities

Let $X_{1}, X_{2}, \ldots, X_{n}$ be zero-mean, jointly Gaussian (not necessarily independent) and identically distributed. As a prerequisite to the derivations that follow, we wish to compute the orthant probability

$$
\mathbb{P}\left(x_{1}>0, X_{2}>0, \ldots, x_{n}>0\right)
$$

The probabilities associated to other sign combinations can easily be deduced from it using the symmetry properties of the Gaussian distribution.
Since the value of the orthant probability does not depend on the common variance of the random variables we may assume without loss of generality that each $X_{i}$ has unit variance: $X_{i} \sim \mathcal{N}(0,1)$. The orthant probability will depend only on the correlation coefficients

$$
\rho_{i, j}=\mathbb{E}\left(X_{i} X_{j}\right) \quad(i \neq j)
$$

## Some lemmas

## Lemma (Quadrant probability of a bivariate normal)

$$
\begin{equation*}
\mathbb{P}\left(X_{1}>0, X_{2}>0\right)=\frac{1}{4}+\frac{\arcsin \rho_{1,2}}{2 \pi} \tag{5}
\end{equation*}
$$

Lemma (Orthant probability of a trivariate normal)

$$
\begin{equation*}
\mathbb{P}\left(X_{1}>0, X_{2}>0, X_{3}>0\right)=\frac{1}{8}+\frac{\arcsin \rho_{1,2}+\arcsin \rho_{2,3}+\arcsin \rho_{1,3}}{4 \pi} \tag{6}
\end{equation*}
$$

Lemma (No closed formula for $n>3$ exists...)

## Main Result: Hadamard Codes

We have $M$ responses bits, so $H(B) \leq M$ bits.
When is it possible to have the maximum value $H(B)=M$ bits?

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## Theorem

$H(B)=M$ implies $M \leq n$.
$H(B)=M=n$ bits if and only if $\mathcal{C}$ is a Hadamard $(n, n)$ code.

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$H(B)=M=n$ bits if and only if $\mathcal{C}$ is a Hadamard $(n, n)$ code.

## Proof.

$H(B)=M$ means that all bits $B_{c}$ are independent, i.e., all $Y_{j}=\sum_{i=}^{n} c_{i} X_{i}^{\prime}$ s are independent (uncorrelated), i.e., all $M$ ( $n$-bit) challenges $c(j)$ are orthogonal.

## Hadamard Codes

$n$ orthogonal binary $\pm 1$ vectors form an Hadamard code:
$n=1 \mathcal{C}=(1), H=1$ bit;
$n=2 \mathcal{C}=\left(\begin{array}{rr}1 & 1 \\ 1 & -1\end{array}\right), H=2$ bits;
$n=3$ No Hadamard code! but any $(3,3)$ code $\equiv\left(\begin{array}{rrr}1 & 1 & 1 \\ -1 & 1 & 1 \\ 1 & -1 & 1\end{array}\right)$
$\quad$ for which $\sum=\frac{1}{3} \mathcal{C C}^{t}=\left(\begin{array}{ccc}1 & 1 / 3 & 1 / 3 \\ 1 / 3 & 1 & -1 / 3 \\ 1 / 3 & -1 / 3 & 1\end{array}\right)$ gives

$$
\begin{aligned}
H= & -6\left(\frac{1}{8}+\frac{\arcsin 1 / 3}{4 \pi}\right) \log \left(\frac{1}{8}+\frac{\arcsin 1 / 3}{4 \pi}\right) \\
& -2\left(\frac{1}{8}-3 \frac{\arcsin 1 / 3}{4 \pi}\right) \log \left(\frac{1}{8}-3 \frac{\arcsin 1 / 3}{4 \pi}\right)
\end{aligned}
$$

$$
\approx 2.875<3 \text { bits }
$$

## Hadamard Codes (cont'd)

$$
\begin{aligned}
& n=4\left(\begin{array}{rrrr}
1 & 1 & 1 & 1 \\
1 & -1 & 1 & -1 \\
1 & 1 & -1 & -1 \\
1 & -1 & -1 & 1
\end{array}\right), H=4 \text { bits } \\
& \mathrm{n}=8\left(\begin{array}{rrrrrrrr}
1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\
1 & -1 & 1 & -1 & 1 & -1 & 1 & -1 \\
1 & 1 & -1 & -1 & 1 & 1 & -1 & -1 \\
1 & -1 & -1 & 1 & 1 & -1 & -1 & 1 \\
1 & 1 & 1 & 1 & -1 & -1 & -1 & -1 \\
1 & -1 & 1 & -1 & -1 & 1 & -1 & 1 \\
1 & 1 & -1 & -1 & -1 & -1 & 1 & 1 \\
1 & -1 & -1 & 1 & -1 & 1 & 1 & -1
\end{array}\right), H=8 \text { bits }
\end{aligned}
$$

## Hadamard Codes (cont'd)

$$
\mathrm{n}=12\left(\begin{array}{rrrrrrrrrrrr}
1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\
1 & 1 & 1 & 1 & 1 & 1 & -1 & -1 & -1 & -1 & -1 & -1 \\
1 & 1 & 1 & -1 & -1 & -1 & 1 & 1 & 1 & -1 & -1 & -1 \\
1 & 1 & -1 & 1 & -1 & -1 & 1 & -1 & -1 & 1 & 1 & -1 \\
1 & 1 & -1 & -1 & 1 & -1 & -1 & 1 & -1 & 1 & -1 & 1 \\
1 & 1 & -1 & -1 & -1 & 1 & -1 & -1 & 1 & -1 & 1 & 1 \\
1 & -1 & -1 & -1 & 1 & 1 & 1 & 1 & -1 & -1 & 1 & -1 \\
1 & -1 & 1 & 1 & -1 & -1 & -1 & 1 & -1 & -1 & 1 & 1 \\
1 & -1 & 1 & -1 & -1 & 1 & 1 & -1 & -1 & 1 & -1 & 1 \\
1 & -1 & -1 & 1 & -1 & 1 & -1 & 1 & 1 & 1 & -1 & -1 \\
1 & -1 & -1 & 1 & 1 & -1 & 1 & -1 & 1 & -1 & -1 & 1 \\
1 & -1 & 1 & -1 & 1 & -1 & -1 & -1 & 1 & 1 & 1 & -1
\end{array}\right)
$$

$$
H=12 \text { bits }
$$

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## Beyond $n$ bits

## $n=1$ element $\Longrightarrow H=1$ bit;

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$$
\begin{aligned}
& n=1 \text { element } \Longrightarrow H=1 \text { bit; } \\
& n=2 \text { elements } \Longrightarrow H=2 \text { bits; }
\end{aligned}
$$

## Beyond $n$ bits

$$
\begin{aligned}
& n=1 \text { element } \Longrightarrow H=1 \text { bit; } \\
& n=2 \text { elements } \Longrightarrow H=2 \text { bits; } \\
& H=M \text { (max entropy = number of challenges) } \Longrightarrow H \leq n \text { bits. }
\end{aligned}
$$

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& H=M \text { (max entropy = number of challenges) } \Longrightarrow H \leq n \text { bits. }
\end{aligned}
$$

Common belief that $n$ elements give at most $n$ bits of entropy (SRAM PUFs, delay PUFs).

Q Can we obtain more than $n$ bits by taking more challenges: $M>n$ ?

## Beyond $n$ bits

$$
\begin{aligned}
& n=1 \text { element } \Longrightarrow H=1 \text { bit; } \\
& n=2 \text { elements } \Longrightarrow H=2 \text { bits; } \\
& H=M \text { (max entropy = number of challenges) } \Longrightarrow H \leq n \text { bits. }
\end{aligned}
$$

Common belief that $n$ elements give at most $n$ bits of entropy (SRAM PUFs, delay PUFs).

Q Can we obtain more than $n$ bits by taking more challenges: $M>n$ ?
A Yes!

$$
n<H<M
$$

For $n$ elements, using $M>n$ challenges, the entropy can increase beyond $n$ bits, albeit strictly $<M$.

## $n=3$ elements

$$
M=1 \mathcal{C}_{1}=\left(\begin{array}{lll}
1 & 1 & 1
\end{array}\right) \text { gives } H=1 \text { bit. }
$$

## $n=3$ elements

$$
\begin{aligned}
M=1 & \mathcal{C}_{1}=\left(\begin{array}{lll}
1 & 1 & 1
\end{array}\right) \text { gives } H=1 \text { bit. } \\
M=2 & \mathcal{C}_{2}=\left(\begin{array}{ll}
1 & 1 \\
1 & 1 \\
1
\end{array}\right) \text { gives } H= \\
& -\left(\frac{1}{2}+\frac{\arcsin 1 / 3}{\pi}\right) \log \left(\frac{1}{4}+\frac{\arcsin 1 / 3}{2 \pi}\right)-\left(\frac{1}{2}-\frac{\arcsin 1 / 3}{\pi}\right) \log \left(\frac{1}{4}-\frac{\arcsin 1 / 3}{2 \pi}\right) \approx 1.966 \text { bits. }
\end{aligned}
$$

## $n=3$ elements

$$
\begin{aligned}
M=1 & \mathcal{C}_{1}=\left(\begin{array}{lll}
1 & 1 & 1
\end{array}\right) \text { gives } H=1 \text { bit. } \\
M=2 & \mathcal{C}_{2}=\left(\begin{array}{ll}
1 & 1 \\
1 & 1
\end{array} \frac{1}{1}\right) \text { gives } H= \\
& -\left(\frac{1}{2}+\frac{\arcsin 1 / 3}{}\right) \log \left(\frac{1}{4}+\frac{\arcsin 1 / 3}{2 \pi}\right)-\left(\frac{1}{2}-\frac{\arcsin 1 / 3}{\pi}\right) \log \left(\frac{1}{4}-\frac{\arcsin 1 / 3}{2 \pi}\right) \approx 1.966 \text { bits. } \\
M=3 & \mathcal{C}_{3}=\left(\begin{array}{cc}
1 & 1 \\
1 & 1 \\
1 & 1 \\
1 & 1
\end{array}\right) \text { gives } H= \\
& -\left(\frac{3}{4}+3 \frac{\arcsin 1 / 3}{2 \pi}\right) \log \left(\frac{1}{8}+\frac{\arcsin 1 / 3}{4 \pi}\right)-\left(\frac{1}{4}-3 \frac{\arcsin 1 / 3}{2 \pi}\right) \log \left(\frac{1}{8}-3 \frac{\arcsin 1 / 3}{4 \pi}\right) \approx 2.875 \text { bits. }
\end{aligned}
$$

## $n=3$ elements

$M=1 \mathcal{C}_{1}=\left(\begin{array}{lll}1 & 1 & 1\end{array}\right)$ gives $H=1 \mathrm{bit}$.
$M=2 \mathcal{C}_{2}=\left(\begin{array}{ccc}1 & 1 & 1 \\ 1 & 1 & -1\end{array}\right)$ gives $H=$
$-\left(\frac{1}{2}+\frac{\arcsin 1 / 3}{\pi}\right) \log \left(\frac{1}{4}+\frac{\arcsin 1 / 3}{2 \pi}\right)-\left(\frac{1}{2}-\frac{\arcsin 2 / 3}{\pi}\right) \log \left(\frac{1}{4}-\frac{\arcsin 1 / 3}{2 \pi}\right) \approx 1.966$ bits.
$M=3 \mathcal{C}_{3}=\left(\begin{array}{rrr}1 & 1 & 1 \\ 1 & 1 & -1 \\ 1 & -1 & 1\end{array}\right)$ gives $H=$
$-\left(\frac{3}{4}+3 \frac{\arcsin 1 / 3}{2 \pi}\right) \log \left(\frac{1}{8}+\frac{\arcsin 1 / 3}{4 \pi}\right)-\left(\frac{1}{4}-3 \frac{\arcsin 1 / 3}{2 \pi}\right) \log \left(\frac{1}{8}-3 \frac{\arcsin 1 / 3}{4 \pi}\right) \approx 2.875$ bits.
$M=4 \mathcal{C}_{3}=\left(\begin{array}{rrr}1 & 1 & 1 \\ 1 & 1 & -1 \\ 1 & -1 & 1 \\ -1 & 1 & 1\end{array}\right)$ gives $\approx 3.666$ bits

## $n=3$ elements

$M=1 \mathcal{C}_{1}=\left(\begin{array}{lll}1 & 1 & 1\end{array}\right)$ gives $H=1 \mathrm{bit}$.
$M=2 \mathcal{C}_{2}=\left(\begin{array}{lll}1 & 1 & 1 \\ 1 & 1 & -1\end{array}\right)$ gives $H=$
$-\left(\frac{1}{2}+\frac{\arcsin 1 / 3}{\pi}\right) \log \left(\frac{1}{4}+\frac{\arcsin 1 / 3}{2 \pi}\right)-\left(\frac{1}{2}-\frac{\arcsin 2 / 3}{\pi}\right) \log \left(\frac{1}{4}-\frac{\arcsin 1 / 3}{2 \pi}\right) \approx 1.966$ bits.
$M=3 \mathcal{C}_{3}=\left(\begin{array}{rrr}1 & 1 & 1 \\ 1 & 1 & -1 \\ 1 & -1 & 1\end{array}\right)$ gives $H=$
$-\left(\frac{3}{4}+3 \frac{\arcsin 1 / 3}{2 \pi}\right) \log \left(\frac{1}{8}+\frac{\arcsin 1 / 3}{4 \pi}\right)-\left(\frac{1}{4}-3 \frac{\arcsin 1 / 3}{2 \pi}\right) \log \left(\frac{1}{8}-3 \frac{\arcsin 1 / 3}{4 \pi}\right) \approx 2.875$ bits.
$M=4 \mathcal{C}_{3}=\left(\begin{array}{rrr}1 & 1 & 1 \\ 1 & 1 & -1 \\ 1 & -1 & 1 \\ -1 & 1 & 1\end{array}\right)$ gives $\approx 3.666$ bits


## $n=4$ elements


$\mathcal{C}_{8}=\left(\begin{array}{rrrr}1 & 1 & 1 & 1 \\ 1 & -1 & 1 & -1 \\ 1 & 1 & -1 & -1 \\ 1 & -1 & -1 & 1 \\ \hline-1 & 1 & 1 & 1 \\ 1 & -1 & 1 & 1 \\ 1 & 1 & -1 & 1 \\ 1 & 1 & 1 & -1\end{array}\right)$
$H=6.251$ bits.

## $n=8$ elements, etc.



## $n=8$ elements, etc.




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## Conclusions and Perspectives

## Conclusions

- $H_{n}=n$ bits of entropy obtained using a Hadamard challenge code;
- $H_{n}>n$ bits of entropy obtained using a challenge code made of several Hadamard "chunks"

Related talk given at ISIT 2016 [RSGD16]:

[CDGB12] Zouha Cherif, Jean-Luc Danger, Sylvain Guilley, and Lilian Bossuet. An easy-to-design PUF based on a single oscillator: The loop PUF.
In 15th Euromicro Conference on Digital System Design, DSD 2012, Çeşme, Izmir, Turkey, September 5-8, 2012, pages 156-162. IEEE Computer Society, 2012.
[PDW89] Marcel J.M. Pelgrom, Aad C.J. Duinmaijer, and Anton P.G. Welbers.
Matching properties of MOS transistors.
IEEE Journal of Solid State Circuits, 24(5):1433-1439, 1989.
DOI: 10.1109/JSSC.1989.572629.
[RSGD16] Olivier Rioul, Patrick Solé, Sylvain Guilley, and Jean-Luc Danger.
On the Entropy of Physically Unclonable Functions.
In ISIT, IEEE International Symposium on Information Theory, July 2016.
Barcelona, Spain.

